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See 37 C.F.R. §§ 1.27 and 1.28.

TOTAL AMOUNT OF PAYMENT (\$) 690.00

## Complete if Known

Application Number  
Filing Date  
First Named Inventor Feng Qian  
Examiner Name  
Group / Art Unit  
Attorney Docket No. 00-088

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## FEE CALCULATION

### 1. BASIC FILING FEE

Large Entity Fee Code (\$)	Small Entity Fee Code (\$)	Fee Description	Fee Paid
101 690	201 345	Utility filing fee	690
106 310	206 155	Design filing fee	
107 480	207 240	Plant filing fee	
108 690	208 345	Reissue filing fee	
114 150	214 75	Provisional filing fee	

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### 2. EXTRA CLAIM FEES

Total Claims	Extra Claims	Fee from below	Fee Paid
6	-20**	0	0
1	-3**	0	0
Multiple Dependent		0	0

\*\*or number previously paid, if greater; For Reissues, see below

Large Entity Fee Code (\$)	Small Entity Fee Code (\$)	Fee Description
103 18	203 9	Claims in excess of 20
102 78	202 39	Independent claims in excess of 3
104 260	204 130	Multiple dependent claim, if not paid
109 78	209 39	** Reissue independent claims over original patent
110 18	210 9	** Reissue claims in excess of 20 and over original patent

SUBTOTAL (2) (\$) 0

## FEE CALCULATION (continued)

### 3. ADDITIONAL FEES

Large Entity Fee Code (\$)	Small Entity Fee Code (\$)	Fee Description	Fee Paid
105 130	205 65	Surcharge - late filing fee or oath	
127 50	227 25	Surcharge - late provisional filing fee or cover sheet	
139 130	139 130	Non-English specification	
147 2,520	147 2,520	For filing a request for reexamination	
112 920*	112 920*	Requesting publication of SIR prior to Examiner action	
113 1,840*	113 1,840*	Requesting publication of SIR after Examiner action	
115 110	215 55	Extension for reply within first month	
116 380	216 190	Extension for reply within second month	
117 870	217 435	Extension for reply within third month	
118 1,360	218 680	Extension for reply within fourth month	
128 1,850	228 925	Extension for reply within fifth month	
119 300	219 150	Notice of Appeal	
120 300	220 150	Filing a brief in support of an appeal	
121 260	221 130	Request for oral hearing	
138 1,510	138 1,510	Petition to institute a public use proceeding	
140 110	240 55	Petition to revive - unavoidable	
141 1,210	241 605	Petition to revive - unintentional	
142 1,210	242 605	Utility issue fee (or reissue)	
143 430	243 215	Design issue fee	
144 580	244 290	Plant issue fee	
122 130	122 130	Petitions to the Commissioner	
123 50	123 50	Petitions related to provisional applications	
126 240	126 240	Submission of Information Disclosure Stmt	
581 40	581 40	Recording each patent assignment per property (times number of properties)	
146 690	246 345	Filing a submission after final rejection (37 CFR § 1.129(a))	
149 690	249 345	For each additional invention to be examined (37 CFR § 1.129(b))	

Other fee (specify) \_\_\_\_\_

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## SUBMITTED BY

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PATENT

Docket: 00-088

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Viterbi Decoder With Adaptive Traceback

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## BACKGROUND OF THE INVENTION

### Field of the Invention

This invention relates to a method and apparatus for decoding  
5 convolutionally encoded signals. Such signals are commonly used in  
communications and recording systems that employ error correction to  
combat signal corruption.

### Description of the Related Art

10 One increasingly popular redundant coding scheme used in wireless  
and other types of communications systems is convolutional coding, in  
which the coding of particular symbol depends upon the value of that symbol  
and the value of a certain number of symbols preceding and succeeding that  
15 particular symbol.

It would be desirable to improve upon conventional maximum  
likelihood sequence estimation ("MLSE") decoders such as Viterbi decoders,  
which are commonly used to decode convolutionally encoded data. If  $\underline{s}_k$  is  
20 a vector that represents an actual transmitted sequence of symbols and  $\underline{r}$  is a  
vector that represents the actual signals received by a mobile, a Viterbi  
decoder effectively tests all of the possible values of  $\underline{s}_k$  and selects the  $\underline{s}_k$  that  
maximizes the summation ("cross correlation")  $\sum_n c[n]r[n] s_k[n] = \underline{s}_k^T \underline{C} \underline{r}$   
where  $c[n]$  defines the channel gain for a transmitted symbol for sample  $n$   
25 (and thus  $C$  is a diagonal matrix with these values). (The  $\underline{s}_k$  that maximizes

the cross correlation is the  $s_k$  that minimizes the “distance” between  $\underline{s}$  and  $s_k$ .) For more details regarding Viterbi decoders, see, for example, “Digital Communications”, John G. Proakis (3d edition 1995).

5        The best  $s_k$  may be represented as a path through nodes in a diagram (known as a trellis), where lines (“transitions”) between nodes in adjacent time steps represent whether an input symbol (information bit) was a 0 or a 1 (for binary coding). The nodes (“states”) in vertical columns represent the values of prior input symbols. A “path” through the trellis therefore  
10 represents a particular sequence of input symbols. Figure 1 shows an example of a trellis with only two paths shown. The time steps indicate a decoder trellis beginning at time  $t=0$  to time  $t=4$ . At each time step, the decision units of the decoder contain the value of the cross correlation of the most likely path to the state of interest. Thus, the decoder trellis contains  
15 multiple paths from an initial state to a given state several time steps later.

         However, only one path has the highest cross correlation and is the most likely path. The most likely path to a particular state within a given time step is found by starting at that particular state at that given time step  
20 and tracing backward (a “traceback”) along the chosen transitions. Information bits that correspond to the transitions along the path are the decoded data. An information bit equal to 0 is shown as a solid line; an information bit equal to 1 is shown as a dotted line. Thus, the path shown in Figure 1 ending at state A at time  $t=0$  corresponds to an information bit  
25 sequence equal to 0100.

A weight is computed for each state at each time, where the weight for a particular state at a particular time corresponds to the likelihood that the encoder was in that state at that time. The weight is equal to the weight of a previous state that transitioned to the current state plus a weight based upon the likelihood of the transition. For example, node 10 at time  $t=4$  has a metric of four which equals the weight of node 12 at time  $t=3$  plus the weight for the transition between those nodes. The node at any given time with the highest weight is the node that ends the most likely path and is therefore the node from which the traceback will occur.

To achieve good noise performance from the decoding process, the traceback length must generally be several times the constraint length (the number of input bits upon which an output depends; this is the length of a shift register that may be used to perform the encoding) of the code. In current code division multiple access ("CDMA"), a traceback length  $L = 5 \cdot K$ , where  $K$  is the constraint length, is frequently taken as the minimum acceptable traceback length.

For a decoder with  $M$  bits of memory,  $M-L$  valid bits can be obtained after each traceback. If an incoming data frame consists of  $N$  information bits (assumed to be greater than  $M$ ), the number of  $L$  bit tracebacks is  $(N-K)/(M-L)$ , which corresponds to a computation requirement of  $M \cdot (N-K)/(M-L)$ , where a computation is assumed to be a one bit traceback. (The one bit traceback is used simply as a "yardstick.") If  $M=L+1$ , this equation reduces

to  $(L+1)*(N-K)$ , which corresponds to the smallest memory and the greatest computational overhead.

5        Instead of performing periodic, partial tracebacks, if the decoder  
memory is sufficiently large, an entire incoming data frame may be stored  
and a full traceback of length  $(N-K)$  performed. Thus the full traceback is  
less computationally intensive than the periodic traceback scheme by a factor  
of  $M/(M-L)$ . However, larger memories (greater  $M$ ) are costly.

10        It would be desirable to implement the partial traceback scheme, with  
its smaller memory requirement, while at the same time decreasing its  
computational overhead.



## SUMMARY OF THE PRESENT INVENTION

These and other needs are met by the present invention, which provides a method and apparatus for decoding transmitted data that has been  
5 generated by encoding information data with a convolutional encoder that generates convolutional codes based on an input sequence of data, where the encoder has a constraint length  $K$  and a rate  $k/n$ .

According to an embodiment of the present invention,  $L-1$  data bits of  
10 a traceback beginning at a time step  $T$  are stored, where  $L$  is the traceback length; these  $L-1$  data bits are the data bits corresponding to the  $L-1$  time steps backwards from time step  $T$ . The maximum likelihood encoder state for time  $T$  is also saved. (The  $L$ -th data bit is the desired data bit as in conventional convolutional decoders.) In a subsequent partial traceback,  
15 preferably beginning at time  $T+1$  that ends at time step  $T$ , the maximum likelihood encoder state for time  $T$  determined from the partial traceback is compared with the stored encoder state for time  $T$ . If they correspond to the same encoder state, the  $L-1$  stored data bits are designated as the last  $L-1$  data bits of the current (partial) traceback.

## **BRIEF DESCRIPTION OF THE DRAWINGS**

Other objects and advantages of the invention will become apparent  
5 upon reading the following detailed description and upon reference to the  
accompanying drawings in which:

Figure 1 is an example of a trellis diagram.

10 Figure 2 represents a digital communications system, in which a  
convolutional decoder including the present invention may be used.

Figure 3 is a flow chart that shows the operation of a possible  
embodiment of a convolutional decoder constructed according to the  
teachings of the present invention.

15

## **DETAILED DESCRIPTION OF THE INVENTION**

The MLSE decoder that is described is a Viterbi decoder but the invention may be applied to any type of MLSE decoder.

5

### **General System Description**

Figure 2 represents a digital communications system 140 comprising a discrete-time channel 142 interposed between an encoder 144 and a decoder 130. Discrete-time channel 142 comprises a modulator 146, a channel 148 and a demodulator 150. An interleaver 145 is interposed between the encoder 144 and the modulator 146. A deinterleaver 151 is interposed between the decoder 130 and the demodulator 150. Channel 148 may be a transmission channel or a storage medium being written to and read from. Interleaver 145 receives a digital output signal from encoder 144 and interleaves this digital output signal over a certain time period, which is usually predetermined and known as a frame. Modulator 146 serves to translate the digital output signal from interleaver 145 into signals suitable for channel 148 and thereafter drives the signals across channel 148.

20

Channel 148 may suffer from interference that corrupts said signals, the interference possibly taking form in any combination of additive noise, cross channel interference, multi-path interference, and channel fading. Demodulator 150 serves to receive the signals from channel 148 while minimizing the interference as much as is practical, and thereafter translate the signals into digital signals for input to deinterleaver 151, which

deinterleaves the digital signal and provides it to decoder 130. Discrete-time channel 142 can thus be viewed as a unit accepting digital input signals and producing possibly corrupted digital output signals although the present invention is not limited to noisy channels.

5

The decoder 130 has a memory 131.

Encoder 144 is a convolutional encoder which serves to add redundancy to input data signal 152. In particular, the encoder 144  
10 comprises a shift register coupled to various arithmetic units (such as modulo 2 adders) that form n bits of output based on the bits in the shift register. For a simple shift register, where each shift of the register causes one bit to be shifted in and out of the register (i.e. each data bit is shifted to the next location in the register) each input bit corresponds to n bits of encoder 144  
15 output and  $1/n$  is known as the rate of the encoder 144. If the shift register has K elements, the K-1 most recent bits input into the encoder correspond to a state. K is known as the constraint length. For binary input data, there are therefore  $2^{K-1}$  possible encoder states.

20 The added redundancy (n bits per information bit) allows for detection and correction of errors that may result from corruption of signals passing across discrete-time channel 142. The error detection and correction is performed by decoder 130.

### Flow Chart of Operation

Figure 3 is a flow chart that details the traceback operation of the decoder 130. At block 200, the decoder 130 receives input data to be decoded. At block 210, the decoder 130 computes the current branch metric.

- 5 This block is known in the art will not be further discussed. At block 220, the decoder 130 selects the surviving state. This block is well known in the art and will not be further discussed. At block 230, the decoder 130 checks whether  $L+K$  decoding cycles (i.e. one decoding cycle per time step) have occurred, where  $L$  is the traceback length and  $K$  is the constraint length. If  
10 not, control passes back to block 210. If so, control passes to block 240, where the decoder 130 stores the starting state of the traceback in the memory 131 (e.g., with reference to Figure 1, the state A at time  $t=4$ ).

- At block 250, the encoder determines whether the traceback to be  
15 performed is the first traceback of the current data frame. If so, control passes to block 260, where a full traceback is performed. For each time step transition within the traceback, the data bit corresponding to the transition between the current state (e.g. the encoder state at time step  $t=4$ ) and the next most recent state (e.g. the state at time step  $t=3$ ) is stored in the memory 131.  
20 Control passes back to block 210. Otherwise, if at least one traceback has already occurred for the data frame, control passes to block 270.

In block 270, the decoder 130 performs a partial traceback; in the preferred embodiment, the partial traceback traces back one time step.

- 25 Control passes to block 280, which compares the finishing state of the partial

5      traceback (e.g. the state at time step  $t=4$  for a traceback beginning at time  
t=5) to the initial state of the most previous traceback (e.g. the traceback that  
began at time step  $t=4$ ). If these quantities are not equal, control passes back  
to block 260, and a “normal” traceback is performed (beginning from the end  
of the partial traceback).

10      If these quantities are equal, control passes to block 290, where the L-  
1 bits stored from the previous traceback are designated as the L-1 final bits  
of the current traceback. In other words, the current traceback bits are taken  
to be the concatenation of (a) the bit corresponding to the first encoder  
transition of the current traceback; with (b) the L-1 final bits of the previous  
15      traceback. The last bit stored from the previous traceback (i.e. the bit  
corresponding to the transition from the third to last encoder state to the  
second to last encoder state of the previous traceback) is designated as the  
desired data bit for the traceback. (The transition from the third to last state  
to the second to last state of the previous traceback corresponds to the  
transition from the second to last state to the last state of the current  
20      traceback.)

## 20      Conclusion

Numerous variations and modifications will become apparent to those  
skilled in the art once the above disclosure is fully appreciated. It is intended  
that the following claims be interpreted to embrace all such variations and  
modifications.

## WHAT IS CLAIMED IS:

1. A method for decoding transmitted data that has been generated by  
encoding a sequence of information data with a convolutional encoder  
5 that generates convolutional codes based on an input sequence of  
information data, the encoder characterized by a constraint length  $K$  and a  
rate  $k/n$ , the method employing a decoder with a memory, the method  
comprising the steps of:
  - (a) storing a first encoder state corresponding to a first time step;
  - 10 (b) beginning at the first time step, performing a first traceback through a  
plurality time steps, thereby determining a first input bit corresponding  
to an encoder state transition from the last time step of the first traceback  
to the second to last time step of the first traceback;
  - (c) storing in the decoder memory a first possible input bit corresponding to  
15 a transition from an encoder state at a third time step within the first  
traceback, to an encoder state at a fourth time step within the first  
traceback, where neither the third nor fourth time steps are the last time  
step;

- (d) performing a second traceback beginning at a second time step and encompassing the first time step, thereby determining a second encoder state at the first time step;
- (e) comparing the first encoder state with the second encoder state;
- 5 (f) if the first encoder state is equal to the second encoder state, designating the first possible input data bit as a decoded data bit corresponding to a transition from the encoder state at the third time step to the encoder state at the fourth time step.
2. The method of claim 1 wherein the second time step is the next time step
- 10 after the first time step.
3. The method of claim 1 wherein the third time step is the next time step after the last time step.
4. The method of claim 1 wherein the second traceback ends at the first time step.
- 15 5. The method of claim 1 wherein the second traceback traces back through one time step.
6. The method of claim 1 further comprising the steps of storing in the decoder memory, for each transition within the first traceback, a possible



input bit corresponding to the transition, thereby storing a plurality of possible input data bits in addition to the first possible input bit.

### ABSTRACT OF THE DISCLOSURE

In a convolutional decoder using partial traceback, L-1 data bits of a traceback beginning at a time step T are stored, where L is the traceback  
5 length; these L-1 data bits are the data bits corresponding to the L-1 time steps backwards from time step T. The maximum likelihood encoder state for time T is also saved. (The L-th data bit is the desired data bit as in conventional convolutional decoders.) In a subsequent partial traceback, preferably beginning at time T+1 that ends at time step T, the maximum  
10 likelihood encoder state for time T determined from the partial traceback is compared with the stored encoder state for time T. If they correspond to the same encoder state, , the L-1 stored data bits are designated as the last L-1 data bits of the current (partial) traceback.

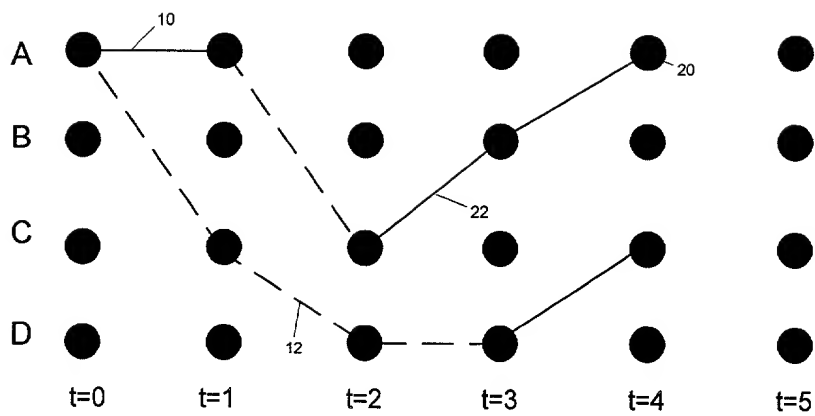


Figure 1

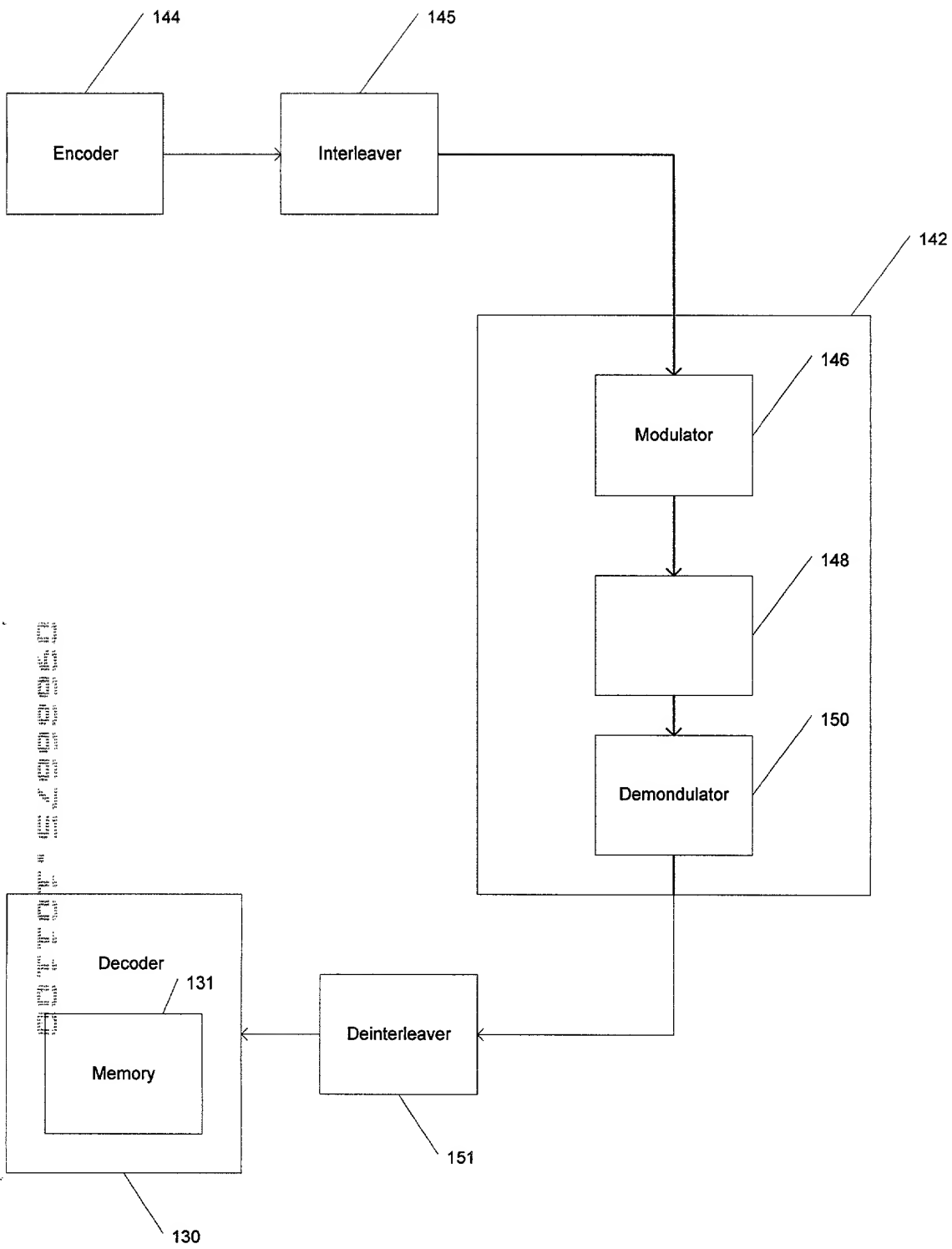


Figure 2

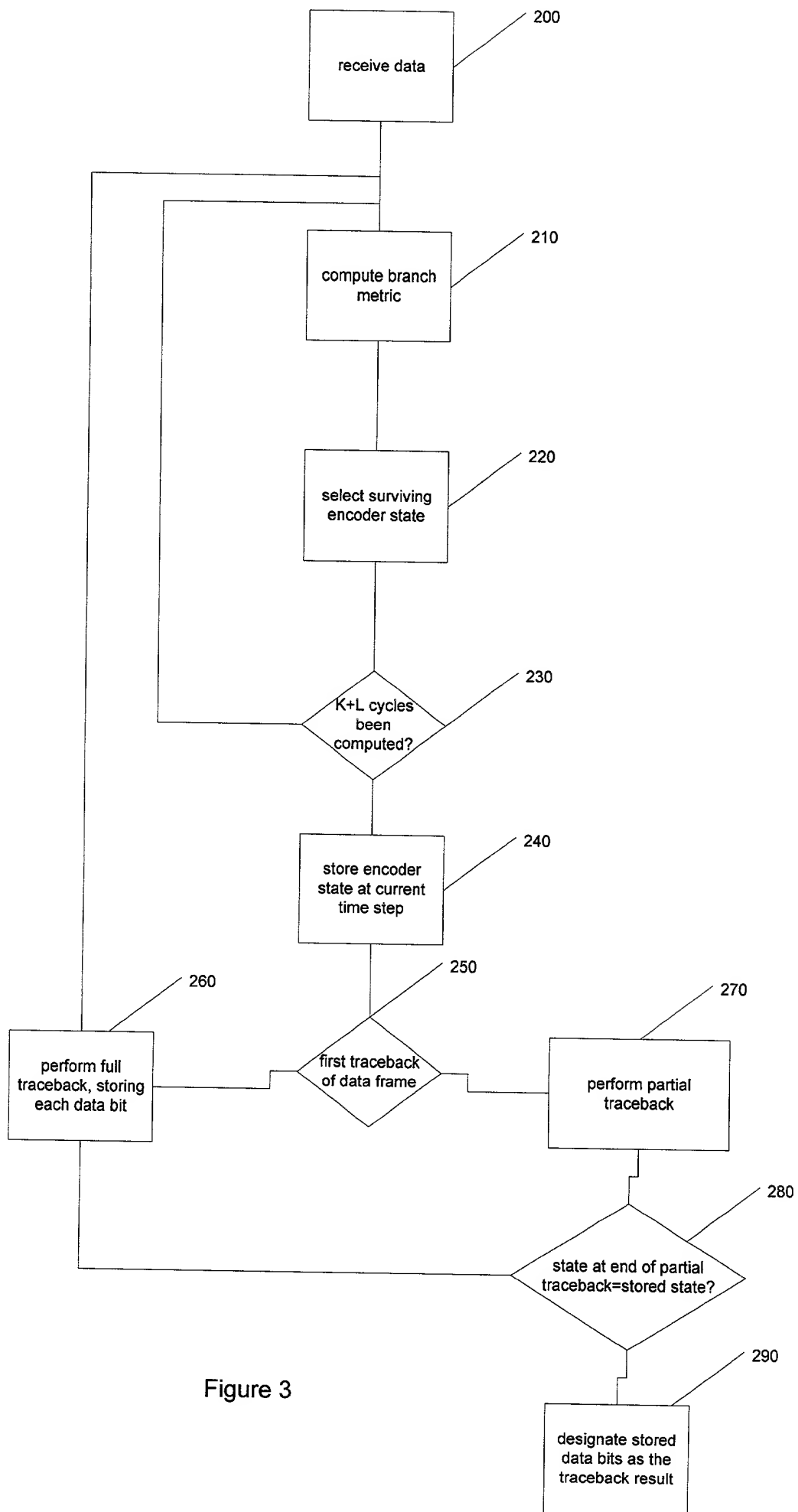


Figure 3

# **Declaration, Power of Attorney, Correspondence Address, and Petition**

**Docket Number :**      **00-088**

## **Declaration**

As a below named inventor, I hereby declare that:

My residence, post office address, and citizenship are as stated below next to my name.

**I believe I am the original, first, and sole inventor of the subject matter which is claimed and for which a patent is sought on the invention entitled:**

**Viterbi Decoder With Adaptive Trace-Back**

**the specification of which is attached hereto.**

I hereby state that I have reviewed and understand the contents of the above identified specification, including the claims, as amended by any amendment referred to above.

I acknowledge the duty to disclose information which is material to the examination of this application in accordance with Title 37, Code of Federal Regulations, § 1.56(a).

I hereby declare that all statements made of my own knowledge are true and that all statements made on information and belief are believed to be true; and further that these statements were made with the knowledge that willful false statements and the like so made are punishable by fine or imprisonment, or both, under Section 1001 of Title 18 of the United States Code and that such willful false statements may jeopardize the validity of the application or any patent issued thereon.

## **Power of Attorney**

I hereby appoint David G. Pursel, Reg. 28,659; Ralph R. Veseli, Reg. 33,807; Bruce R. Hopenfeld, Reg. 39,714; Gary Edward Ross, Reg. 29,431; Lloyd E. Dakin, Reg. 38,423; and Sandeep Jaggi, Reg. 43,331; as my attorneys with full power of substitution and revocation, to prosecute this application and to transact all business in the U.S. Patent and Trademark Office connected therewith and before competent international authorities.

## **Correspondence Address**

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Intellectual Property Services Group  
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Milpitas, CA 95035  
Phone: (408) 433-8708  
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## **Petition**

Wherefore I pray that Letters Patent be granted to me for the invention or discovery described and claimed in the foregoing specification and claims, and I hereby subscribe my name to the foregoing specification and claims, declaration, power of attorney, and this petition.

[illegible]

**First Name** Feng

**Last Name:** Qian

**Inventor's Signature:**

**Date**

8/14/2000

**Address** 25851 Majorca Way

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